

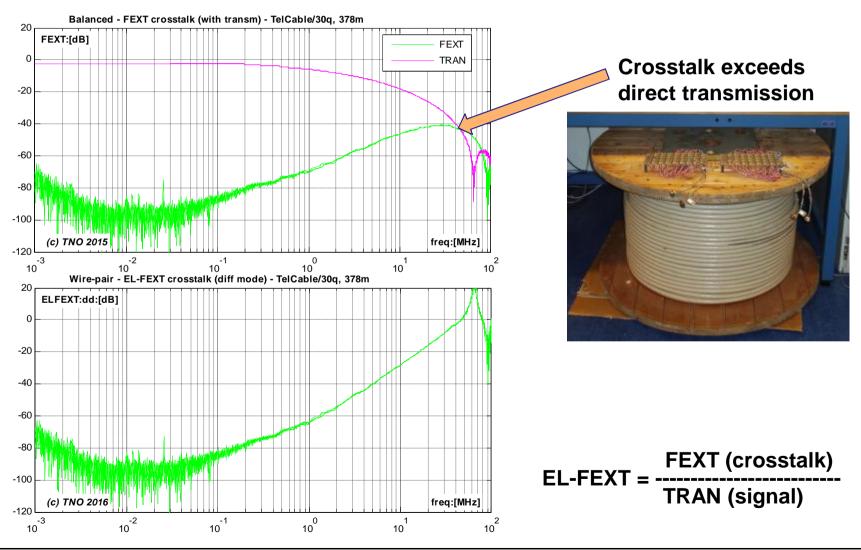
Understanding the dual-slope effect in crosstalk (EL-FEXT)

What causes this second order crosstalk effect in quad cables?

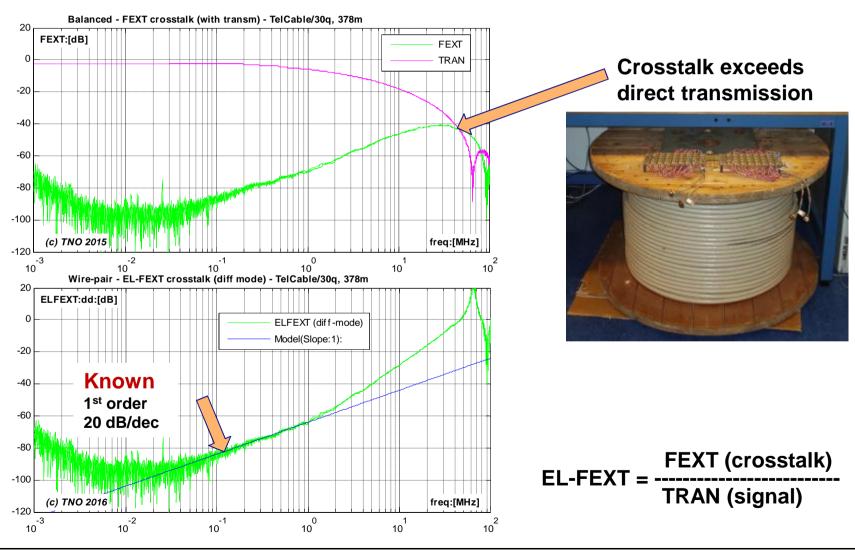
Rob F.M. van den Brink - TNO



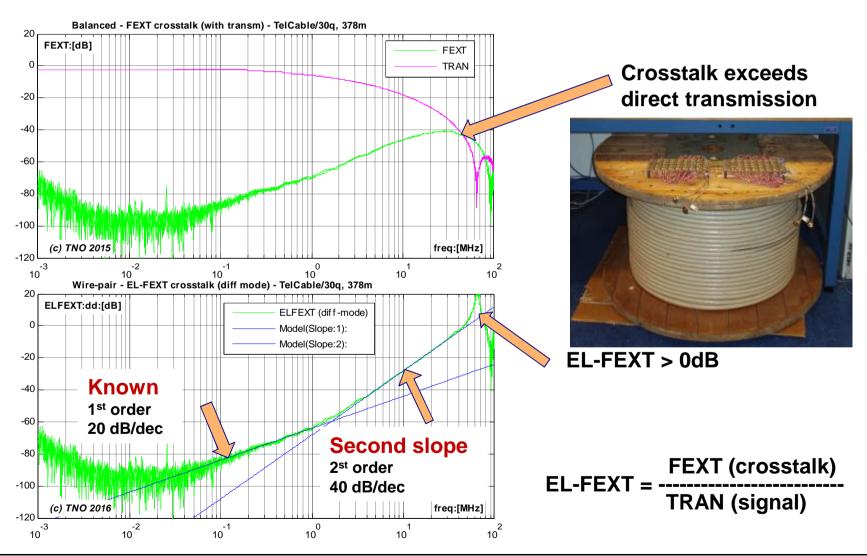




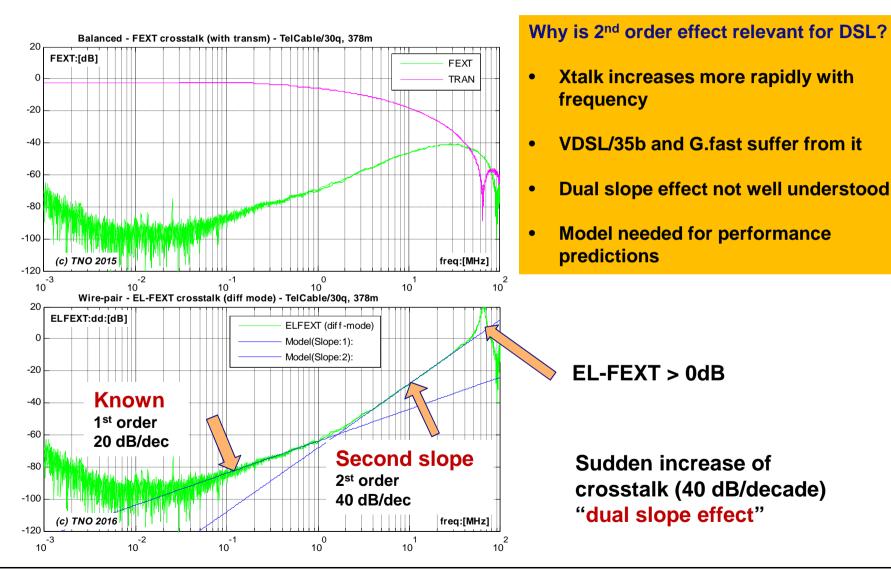








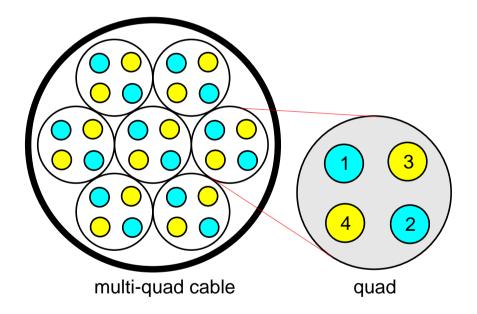






How behaves this dual-slope effect in other cables?

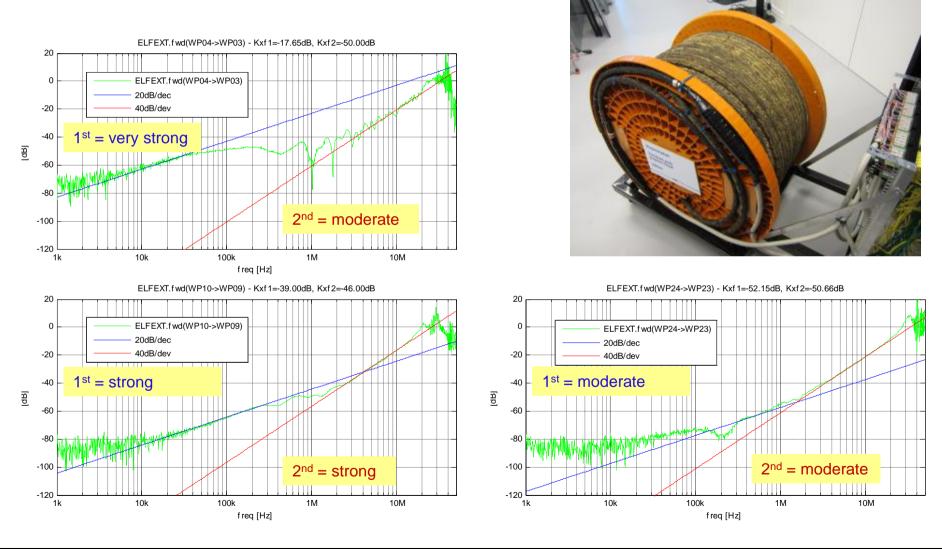
- 1st and 2nd order slopes are caused by independent mechanisms
- They scale differently with the cable length





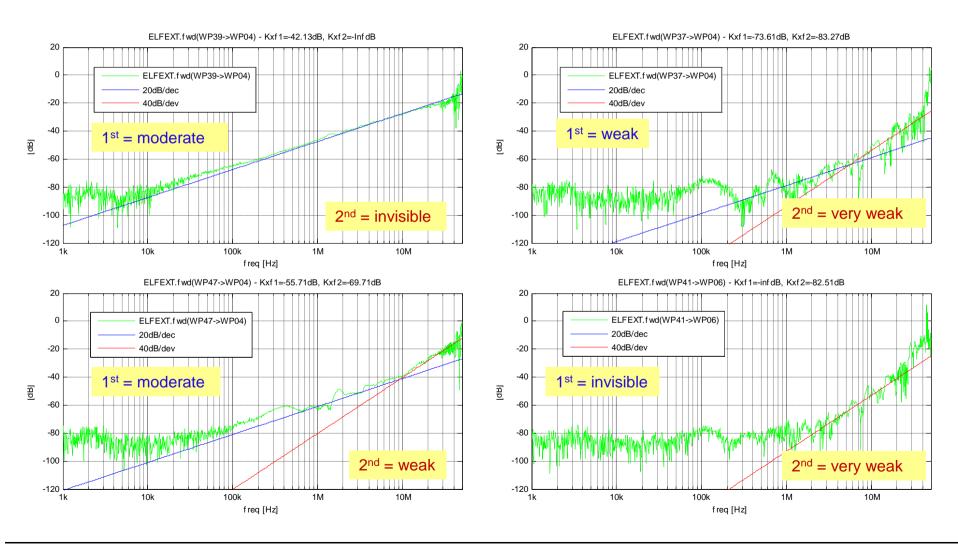


Other example: 280m, GPLK (paper insulated), drum, in-quad





Other example: 280m, GPLK (paper insulated), drum: between quads





Summary of observations so far

	In-quad	Between quads
1st order crosstalk	Moderate à very strong	Invisibleà moderate
2 nd order crosstalk	Moderate à strong	Invisibleà weak

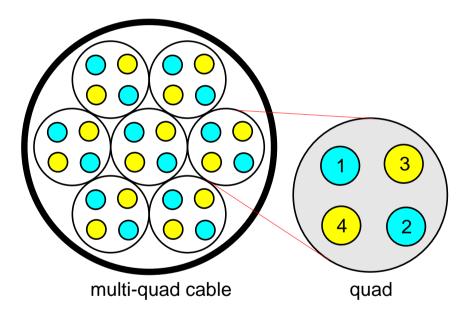
Observations so far:

- Crosstalk with mix of 1st and 2nd order effect
- Crosstalk with 1st order only (2nd order invisible)
- Crosstalk with 2nd order only (1st order invisible)
- 2nd order effect observed in both in-quad and between quads
- In-quad crosstalk more pronounced
 - à Dual slope more visible
 - à Lets focus on in-quad crosstalk to gain understanding



Cause of the first order effect (well known)

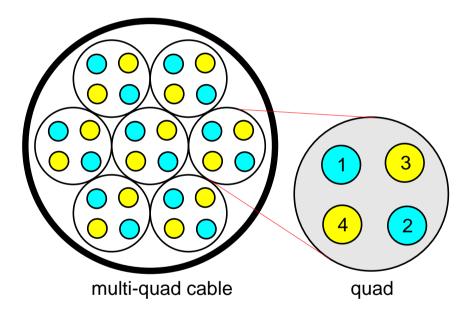






Cause of the first order effect (well known)





perfect square capacitive balance no crosstalk

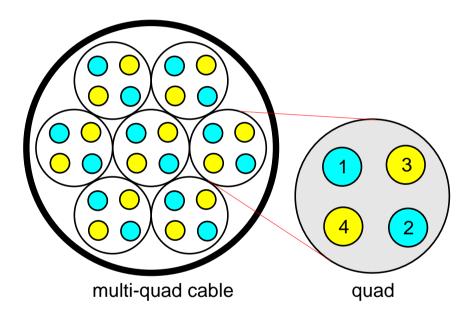


circuit equivalent

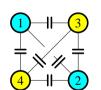


Cause of the first order effect (well known)





perfect square capacitive balance no crosstalk



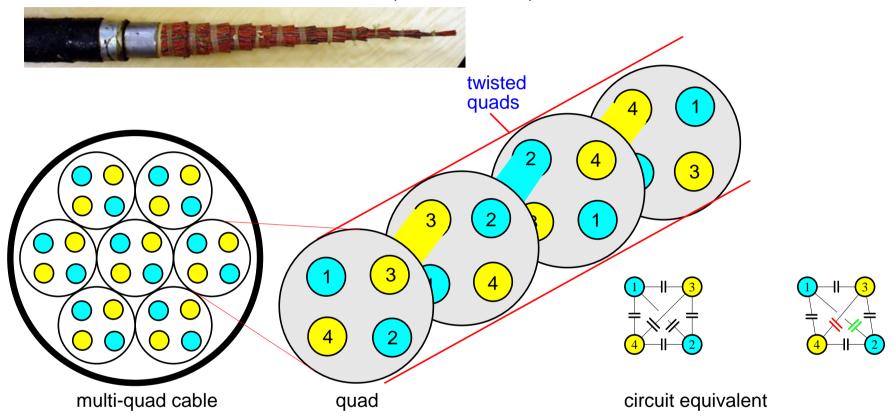
circuit equivalent

random variations small unbalance crosstalk





Cause of the first order effect (well known)



Twisting can reduce crosstalk

- Does <u>not</u> reduce 1st order effect <u>within</u> a quad
- Reduces 1st order crosstalk between different quads
- Different quads should have different twist lengths



Cause of the first order effect (well known)



First order effect of EL-FEXT (well known):

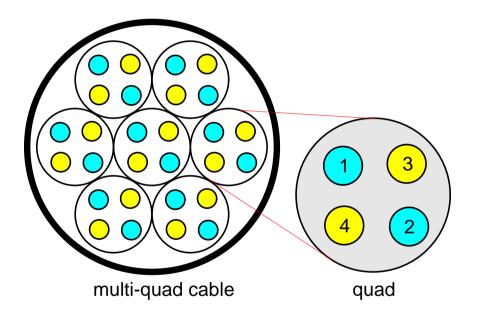
- Random perturbation of balance by imperfect geometry
- Scales proportionally with frequency (20 dB/decade)
- Scales with the root of the cable length (\sqrt{L})

Bell technical staff, *Transmission*Systems for Communications,
Bell Telephone Laboratories,
1969 (1st ed 1954)



Cause of the second order effect





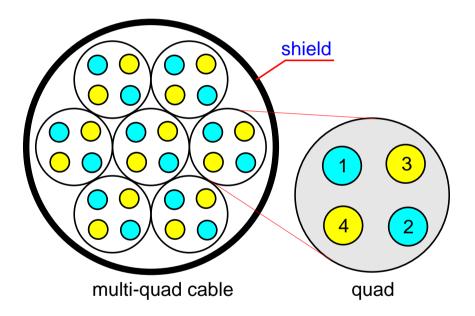


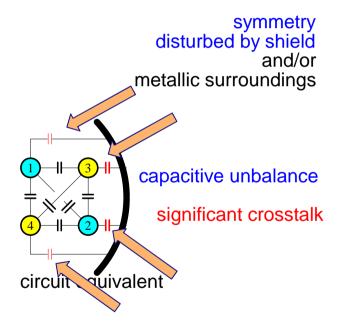
circuit equivalent



Cause of the second order effect

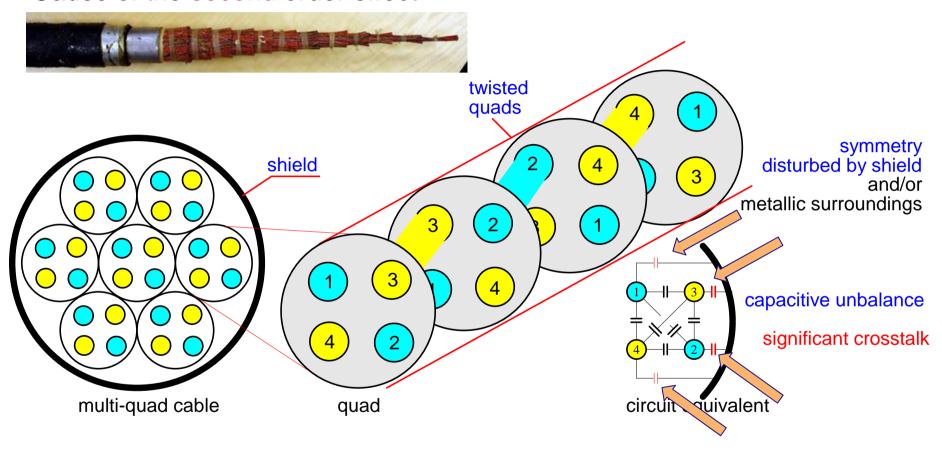








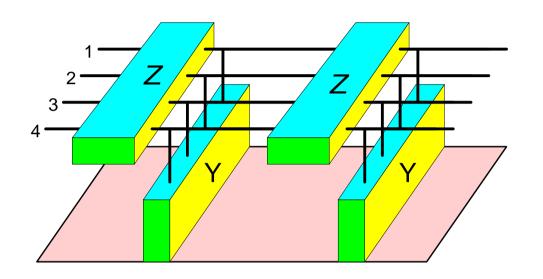
Cause of the second order effect



Twisting a quad restores capacitive balance only on <u>average</u>
Significant crosstalk reduction within the quad à residual crosstalk
Tighter twisting of a quad à lowers residual crosstalk (= 2nd order effect)

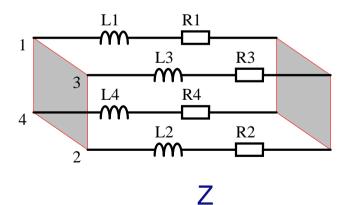


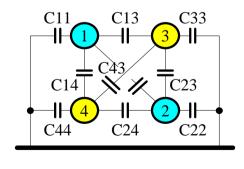
Four wires and a shield = eight port



$$\frac{\partial}{\partial z}\mathbf{U}(z) = -\mathbf{Z}_{s} \cdot \mathbf{I}(z)$$

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$$\frac{\partial}{\partial z} \mathbf{I}(z) = -\mathbf{Y}_p \cdot \mathbf{U}(z)$$

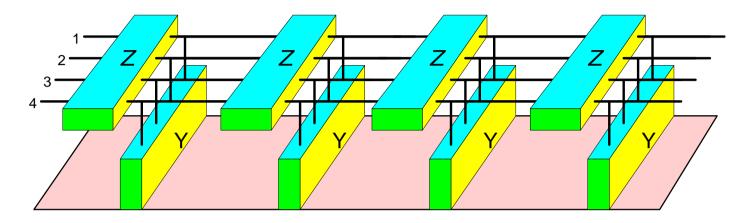




Clayton R. Paul, Analysis of multiconductor transmission lines, IEEE press, 2008



"Infinite" cascade à eight port cable model

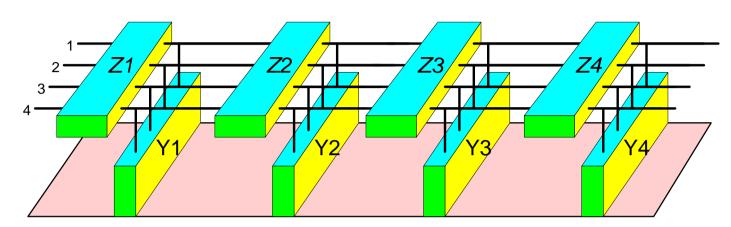


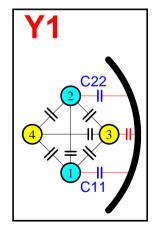
With only 9 values (for R,L,C) you can do a very good cable simulation

.... but the dual slope effect will not be visible

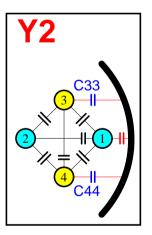


"Infinite" cascade à eight port cable model à adding a twist

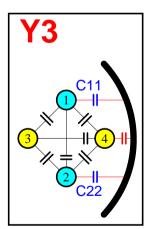




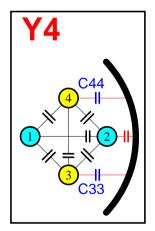
C11=Cc C22=Cc C33=Cc+dC C44=Cc-dC



C11=Cc+dC C22=Cc-dC C33=Cc C44=Cc



C11=Cc C22=Cc C33=Cc-dC C44=Cc+dC



C11=Cc-dC C22=Cc+dC C33=Cc C44=Cc Cascade of piecewise uniform segments can add the twist to the cable

4 sections/twist?

16 sections/twist?

64 sections/twist?

$$C_{11} = C_c / (1 + \Delta_c \cdot \sin(2p \cdot k/n \cdot \Delta L))$$

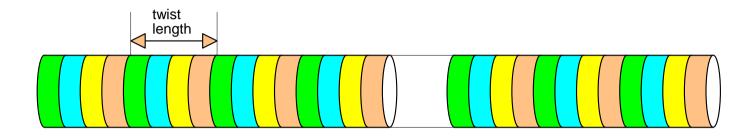
$$C_{22} = C_c / (1 - \Delta_c \cdot \sin(2p \cdot k / n \cdot \Delta L))$$

$$C_{33} = C_c / (1 + \Delta_c \cdot \cos(2p \cdot k/n \cdot \Delta L))$$

$$C_{44} = C_c / (1 - \Delta_c \cdot \cos(2p \cdot k/n \cdot \Delta L))$$



"Infinite" cascade à eight port cable model



Twist length = 3 cm
Cable length = 300m
Model with 4 segments per twist

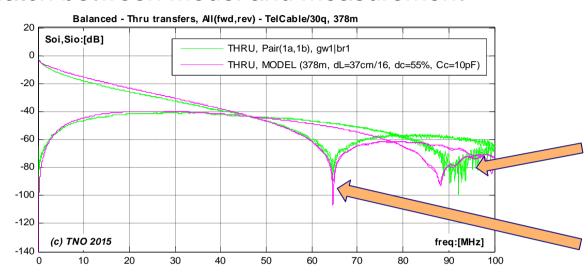
- à 10000 twists in a cable
- à 40000 segments in a cable
- à And they all should go in the cascade

That was the theory,

But is it also realistic, and is it good enough?



Match between model and measurement



Very good match between model and measurement

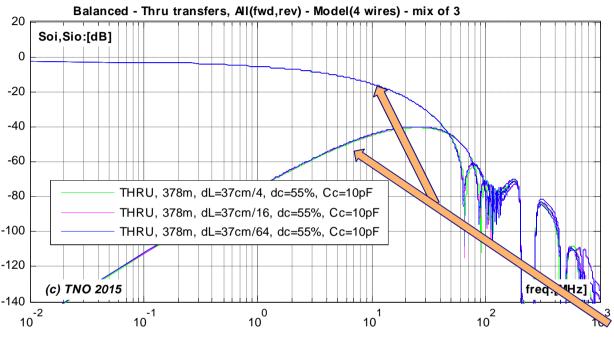
This error is because we ignored that insulator has a frequency dependency

Even the dip is well modelled





How many segments per twist are needed for a good match?



4 segments/twist appears to be enough!

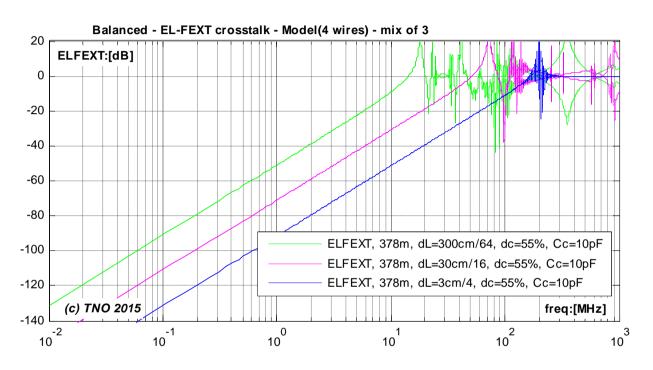
Three different simulation runs ...

With different segments/twist 4 segments/twist 16 segments/twist 64 segments/twist

You cannot see the difference!



Dual slope effect, as a function of the twist length



Three different simulation runs

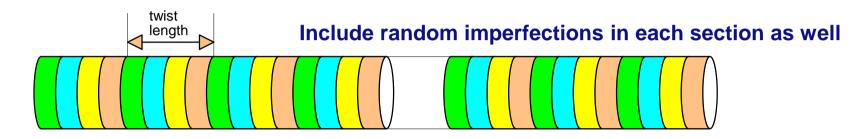
With different twist lengths 300 cm 30 cm 3 cm

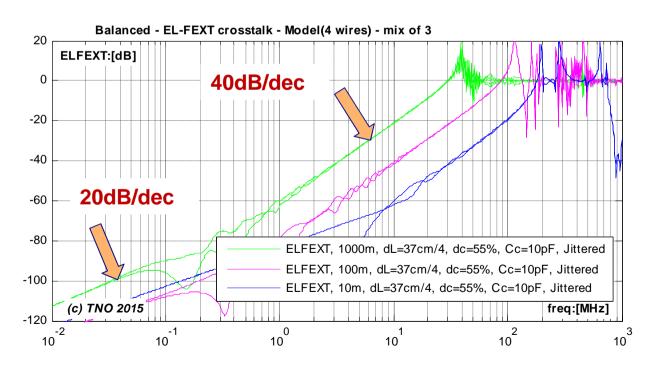
EL-FEXT has 40 dB slope (second order effect)
Even when all random perturbations (first order effect) is absent

An decrease of the twist length by a factor 10 ... decreases the EL-FEXT by 20 dB (second order effect)



Fair prediction as function of frequency and cable length for both the first <u>and</u> second order effects!





Three different simulation runs

Changing the loop length

10m 100m 1000m



Cause of the first and second order effect (well known)



First order effect of EL-FEXT (well known):

- Random perturbation of balance by imperfect geometry
- Scales proportionally with frequency (20 dB/decade)
- Scales with the root of the cable length (√L)
 (only on a statistical sense)

Second order effect of EL-FEXT

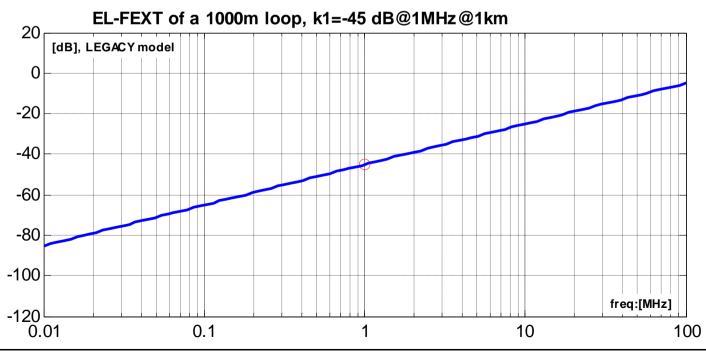
- Deterministic perturbation of balance by metallic surroundings
- Scales proportionally with square of frequency (40 dB/decade)
- Scales with the cable length (L)



Fair prediction as function of frequency and cable length

Legacy system model for EL-FEXT (first order only, ETSI TR 101 830-2)

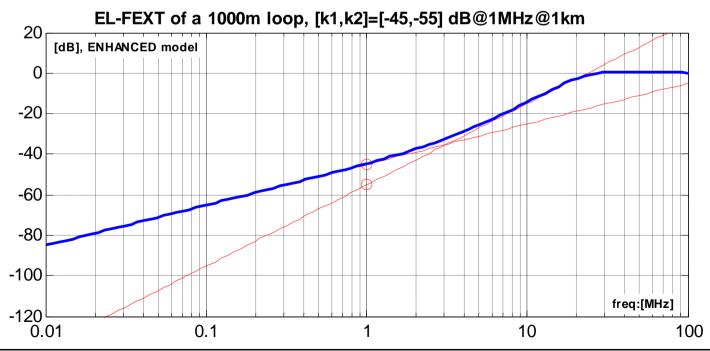
- First order slope of 20 dB/decade
- Scales with the root of cable length
- No second order slope
- Infinite high EL-FEXT at high frequencies





Fair prediction as function of frequency and cable length

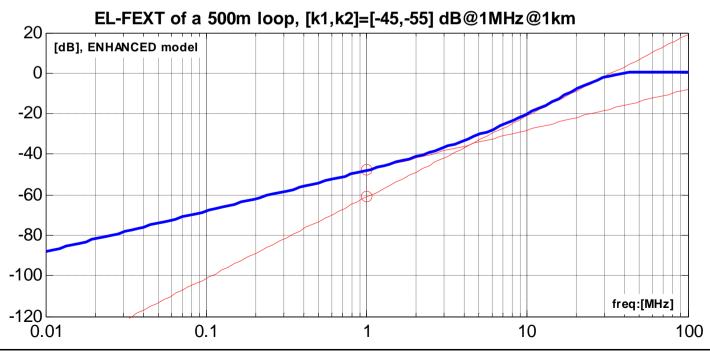
- First order slope of 20 dB/decade
- Second order slope of 40 dB/decade
- Does not exceed 0dB for high frequencies
- First order slope scales with the root of cable length
- Second order slope scales linear with the cable length





Fair prediction as function of frequency and cable length

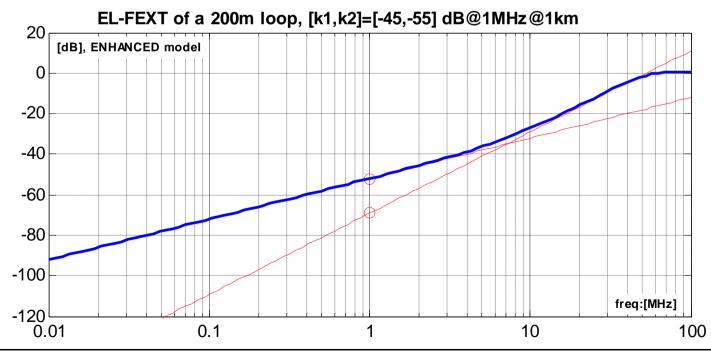
- First order slope of 20 dB/decade
- Second order slope of 40 dB/decade
- Does not exceed 0dB for high frequencies
- First order slope scales with the root of cable length
- Second order slope scales linear with the cable length





Fair prediction as function of frequency and cable length

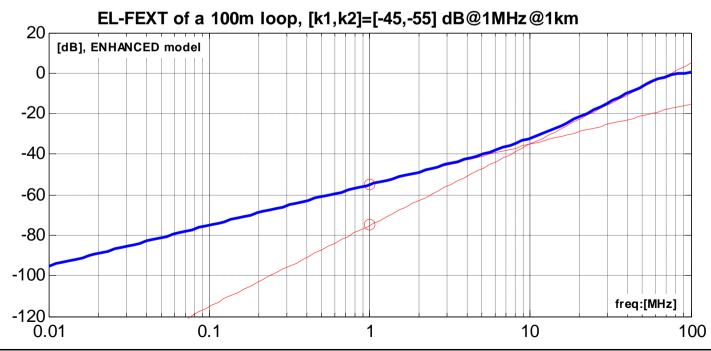
- First order slope of 20 dB/decade
- Second order slope of 40 dB/decade
- Does not exceed 0dB for high frequencies
- First order slope scales with the root of cable length
- Second order slope scales linear with the cable length





Fair prediction as function of frequency and cable length

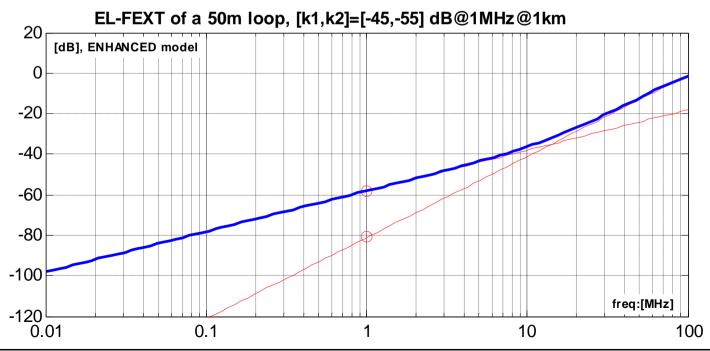
- First order slope of 20 dB/decade
- Second order slope of 40 dB/decade
- Does not exceed 0dB for high frequencies
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Fair prediction as function of frequency and cable length

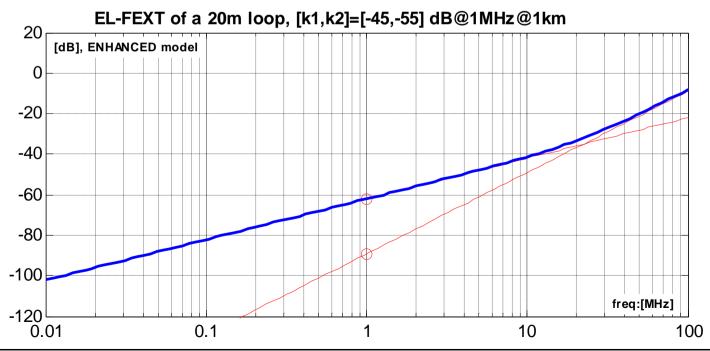
- First order slope of 20 dB/decade
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Fair prediction as function of frequency and cable length

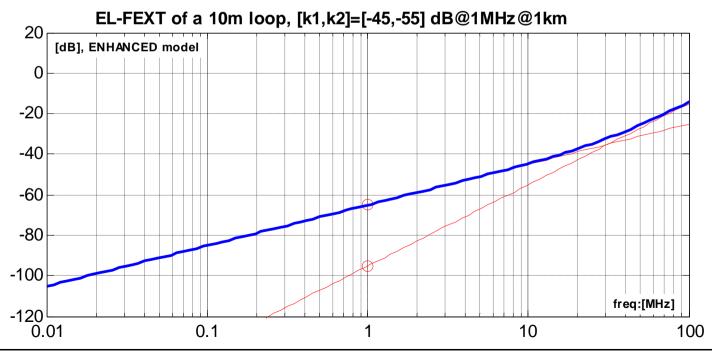
- First order slope of 20 dB/decade
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- Does not exceed 0dB for high frequencies
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- Second order slope scales linear with the cable length





Fair prediction as function of frequency and cable length

- First order slope of 20 dB/decade
- Second order slope of 40 dB/decade
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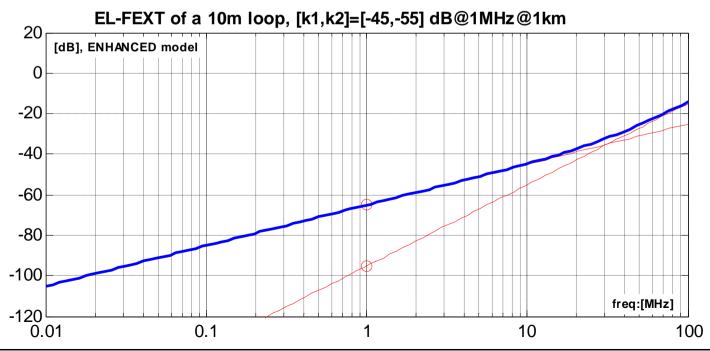


Fair prediction as function of frequency and cable length

Enhanced system model for EL-FEXT (first + second order)

- First order slope of 20 dB/decade
- Second order slope of 40 dB/decade
- Does not exceed 0dB for high frequencies
- First order slope scales with the root of cable length
- Second order slope scales linear with the cable length

A second order highpass filter curve meets all these requirements



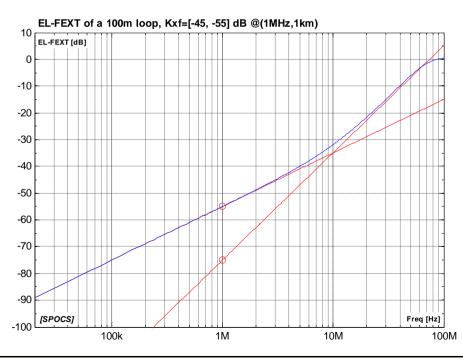


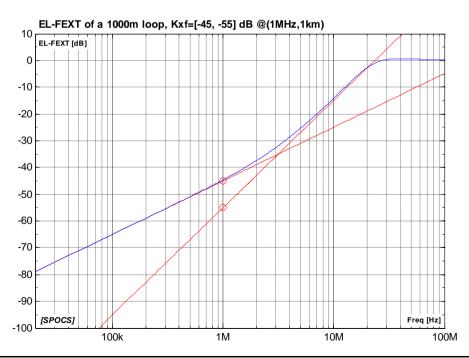
Fair prediction as function of frequency and cable length

Enhanced model for EL-FEXT (first + second order)

$$H_{ELFEXT}(jw,L) = \frac{k_1(L) \cdot \binom{jw}{w_0} + k_2(L) \cdot \binom{jw}{w_0}^2}{1 + (k_1(L) + \sqrt{k_2(L)}) \cdot \binom{jw}{w_0} + k_2(L) \cdot \binom{jw}{w_0}^2}$$
e.g.
$$K_{XF1} = -45 \text{dB (@1MHz, 1km)}$$
$$K_{XF2} = -55 \text{dB (@1MHz, 1km)}$$

where $k_1(L) = K_{YE_1} \cdot \sqrt{(L/L_0)}$, and $k_2(L) = K_{YE_2} \cdot (L/L_0)$







5. Conclusions

Discussed the dual slope effect in crosstalk (EL-FEXT)

- EL-FEXT increases more rapidly at higher frequencies (40 dB/decade):
 - Dual slope effect in EL-FEXT raised in ITU, Feb 2012 (2012-04-4A-038)
 - Confirmed by many measurements in different labs.
 - Has an impact on both G.fast and VDSL performance (above a few MHz)
 - Vector engines should cope with it.
- Achieved a good understanding on the origin of this effect:
 - Derived via an advanced brute force multi-port model of a cable (L,C,R)
 - 2nd order effect due to *deterministic* interaction of twist in quads and their metallic surroundings
 - 1st order effect due to *random* perturbation in geometry
 - Both effects are independent from each other
 - Both effects scale differently with cable length
 - Details in IEEE paper (submitted in Jan & May 2016)
- Proposed a simple FEXT model for performance simulations:
 - Simple extension to legacy crosstalk model (ETSI)
 - Handles magnitude (both slopes) and length scaling quite well
 - Convenient for performance simulations on VDSL & G.fast
 - Contributed to ITU, April 2016 (2016-04-Q4-021), recommended for BBF TR-285

"Dual slope" effect in EL-FEXT is now well understood





